

Signal Analysis & Communication ECE 355

Ch. 2.3: Properties of LTI Systems

Lecture 4

27-09-2023



Ch. 2.3: PROPERTIES OF LTI SYSTEMS

- Recall that a (CT/DT) LTI system is completely characterized by an impulse response, $h(t)/h[n]$.

(that's not the case with non-linear & time variant systems)

- LTI systems possess a number of useful properties not possessed by other systems.

① COMMUTATIVE PROPERTY (Convolution is commutative)

CT: $x(t) * h(t) = h(t) * x(t)$

Proof

$$\text{LHS } x(t) * h(t) = \int_{-\infty}^{\infty} x(\tau) h(t-\tau) d\tau$$

$$\text{let } s = (t-\tau)$$

$$ds = -d\tau; \text{ limits: } \infty \rightarrow -\infty \quad \left| \begin{array}{l} \Rightarrow ds \text{ with} \\ \text{limits: } -\infty \rightarrow \infty \end{array} \right.$$

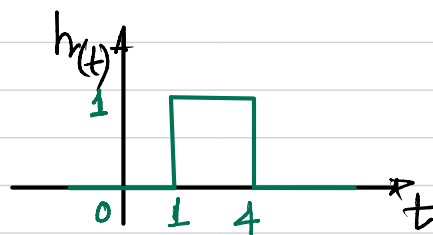
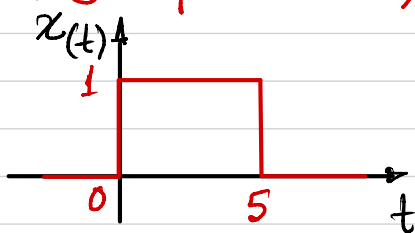
$$\therefore x(t) * h(t) = \int_{-\infty}^{\infty} x(t-s) h(s) ds$$

$$= \int_{-\infty}^{\infty} h(s) x(t-s) ds = \text{RHS}$$

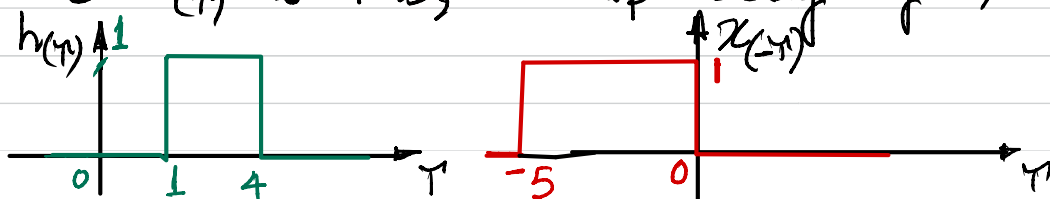
DT: Similar.

Example ①

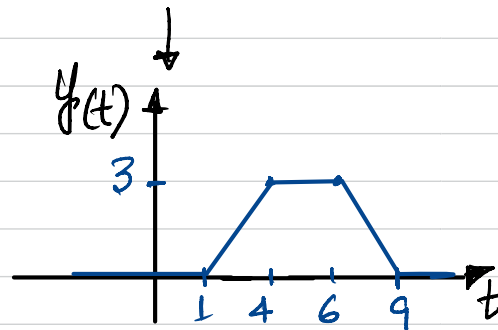
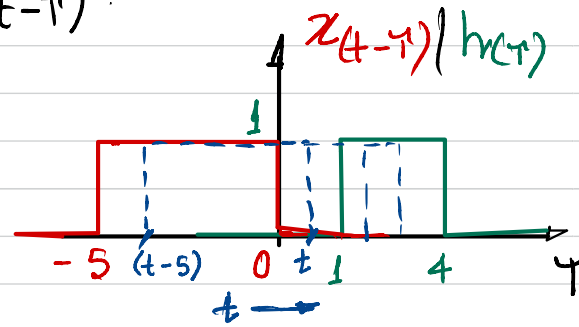
(Ex. ② of prev. lecture)



Take $h(\tau)$ as it is, & flip & delay (by 't') $x(\tau)$ here.



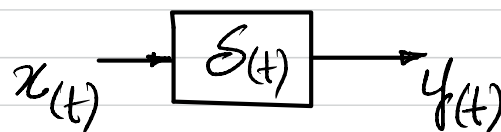
Slide $x(-\tau)$ by ' t ' for different regions of overlap of $h(\tau)$ & $x(t-\tau)$.



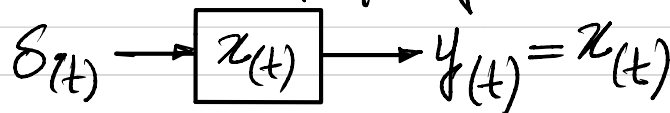
Same result as previous.

Example ②

$$h(t) = \delta(t)$$



By commutative property.



We know when
 $x(t) = \delta(t)$
 $y(t) = h(t)$

CALLED IDENTITY SYSTEM

Basically,

$$\int_{-\infty}^{\infty} x(\tau) \delta(t-\tau) d\tau = x(t)$$

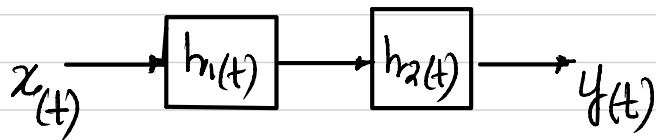
'sifting' of
unit impulse.

$$= \int_{-\infty}^{\infty} \delta(\tau) x(t-\tau) d\tau = x(t)$$

↑
exists at $\tau=0$

② ASSOCIATIVE PROPERTY (Convolution is associative)

CT:



$$y(t) = [x(t) * h_1(t)] * h_2(t) = x(t) * [h_1(t) * h_2(t)]$$

Proof:

$$\text{LHS} = \left[\int_{-\infty}^{\infty} x(\tau) h_1(t-\tau) d\tau \right] * h_2(t)$$

$$= \int_{-\infty}^{\infty} \left[\int_{-\infty}^{\infty} x(\tau) h_1(s-\tau) d\tau \right] h_2(t-s) ds$$

$$= \int_{-\infty}^{\infty} x(\tau) \left[\int_{-\infty}^{\infty} h_1(s-\tau) h_2(t-s) ds \right] d\tau \quad \text{--- (1)}$$

$$\text{RHS} = \int_{-\infty}^{\infty} x(\tau) [h_1 * h_2]_{(t-\tau)} d\tau$$

$$= \int_{-\infty}^{\infty} x(\tau) \left[\int_{-\infty}^{\infty} h_1(\tau') h_2(t-\tau-\tau') d\tau' \right] d\tau$$

NOTE: $y(t) = (x * h)_{(t)} = x_{(t)} * h_{(t)} = \int_{-\infty}^{\infty} x(\tau) h_{(t-\tau)} d\tau$

$$y_{(t-1)} = \underbrace{(x * h)_{(t-1)}}_{\neq x_{(t-1)} * h_{(t-1)}} = \int_{-\infty}^{\infty} x(\tau) \cdot h_{(t-1)-\tau} d\tau$$

Now let $\tau' = s - \tau$, $d\tau' = ds$; limits: same

$$= \int_{-\infty}^{\infty} x(\tau) \left[\int_{-\infty}^{\infty} h_1(s-\tau) h_2(t-s) ds \right] d\tau \quad \text{--- (2)}$$

System in Series

Eqn (1) = Eqn (2)



NOTE: 1. We can simply write $x(t) * h_1(t) * h_2(t)$

2. Commutative + Associative

$$x(t) * h_1(t) * h_2(t) = x(t) * h_2(t) * h_1(t)$$

3. Extend to any number of systems in series.

$$x(t) * h_1(t) * h_2(t) * \dots * h_n(t) = y(t)$$

DT: Similar

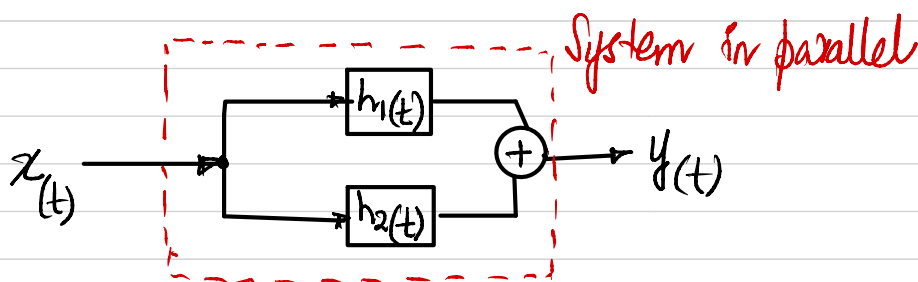
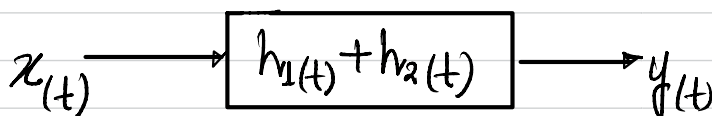
③ DISTRIBUTIVE PROPERTY (Convolution is distributive)

CT: $x(t) * [h_1(t) + h_2(t)] = x(t) * h_1(t) + x(t) * h_2(t)$

Proof

I. Using LHS = RHS, $\int_{-\infty}^{\infty} x(\tau) \dots d\tau = \dots$

II. View $x(t)$ as sig. & $[h_1(t) + h_2(t)]$ as sys.



Example

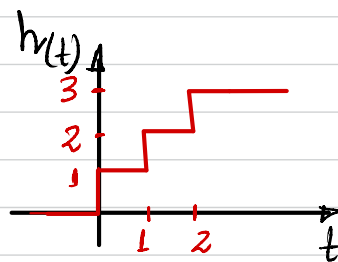
$$h(t) = u(t) + u(t-1) + u(t-2)$$

$$x(t) = e^{-at} u(t)$$

Find $y(t) = x(t) * h(t)$

$$x(t) * u(t) = \frac{1}{a} (1 - e^{-at}) u(t) \quad (\text{Example in Lecture 8})$$

$$x(t) * u(t-1) = \frac{1}{a} (1 - e^{-a(t-1)}) u(t-1)$$



$$x(t) * u_{(t-2)} = \frac{1}{a} (1 - e^{-a(t-2)}) u_{(t-2)}$$

Therefore,

$$y(t) = x(t) * h(t) = \frac{1}{a} [(1 - e^{-at}) u_{(t)} + (1 - e^{-a(t-1)}) u_{(t-1)}]$$

